

# MIC49300

3.0A Low Voltage LDO Regulator w/Dual Input Voltages

### **General Description**

The MIC49300 is a high-bandwidth, low-dropout, 3.0A voltage regulator ideal for powering core voltages of low-power microprocessors. The MIC49300 implements a dual supply configuration allowing for very low output impedance and very fast transient response.

The MIC49300 requires a bias input supply and a main input supply, allowing for ultra-low input voltages on the main supply rail. The input supply operates from 1.4V to 6.5V and the bias supply requires between 3V and 6.5V for proper operation. The MIC49300 offers fixed output voltages from 0.9V to 1.8V and adjustable output voltages down to 0.9V.

The MIC49300 requires a minimum of output capacitance for stability, working optimally with small ceramic capacitors.

The MIC49300 is available in a 5-pin S-Pak. It operates over a junction temperature range of  $-40^{\circ}$ C to  $+125^{\circ}$ C.

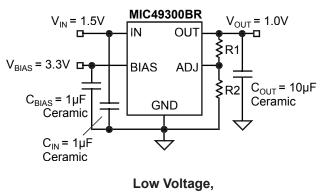
### Features

- Input Voltage Range: V<sub>IN</sub>: 1.4V to 6.5V V<sub>BIAS</sub>: 3.0V to 6.5V
- Stable with 1µF ceramic capacitor
- ±1% initial tolerance
- Maximum dropout voltage (V<sub>IN</sub>-V<sub>OUT</sub>) of 500mV over temperature
- Adjustable output voltage down to 0.9V
- Ultra fast transient response (Up to 10MHz bandwidth)
- · Excellent line and load regulation specifications
- · Logic controlled shutdown option
- · Thermal shutdown and current limit protection
- Power S-Pak package
- Junction temperature range: –40°C to 125°C

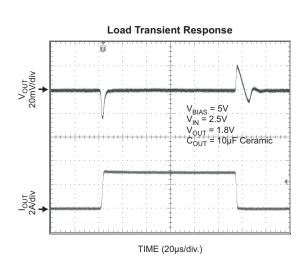
### **Applications**

- Graphics processors
- PC Add-In Cards
- · Microprocessor core voltage supply
- Low voltage digital ICs
- · High Efficiency Linear power supplies
- SMPS post regulators

### **Typical Application**



Fast Transient Response Regulator

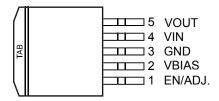


## **Ordering Information**

Part Number	<b>RoHS</b> Compliant*	Output Current	Voltage	Temperature Range	Package
MIC49300-0.9BR	MIC49300-0.9WR*	3A	0.9V	–40°C to +125°C	S-Pak-5
MIC49300-1.2BR	MIC49300-1.2WR*	3A	1.2V	–40°C to +125°C	S-Pak-5
MIC49300-1.5BR	MIC49300-1.5WR*	3A	1.5V	–40°C to +125°C	S-Pak-5
MIC49300-1.8BR	MIC49300-1.8WR*	3A	1.8V	–40°C to +125°C	S-Pak-5
MIC49300BR	MIC49300WR*	3A	ADJ.	–40°C to +125°C	S-Pak-5

Other voltages available. Contact Micrel for details. \*RoHS compliant with 'high-melting solder' exemption.

## **Pin Configuration**



5-Lead S-Pak (R)

## **Pin Description**

Pin Number	Pin Name	Pin Function
1	Enable	Enable (Input): CMOS compatible input. Logic high = enable, logic low = shutdown.
	ADJ.	Adjustable regulator feedback input. Connect to resistor voltage divider.
4	VIN	Input voltage which supplies current to the output power device.
5	VOUT	Regulator Output.
2	VBIAS	Input Bias Voltage for powering all circuitry on the regulator with the excep- tion of the output power device.
3	GND	Ground (TAB is connected to ground on S-Pak).

### Absolute Maximum Ratings (Note 1)

Supply Voltage (V <sub>IN</sub> )	
Bias Supply Voltage (V <sub>BIAS</sub> )	8V
Enable Input Voltage (V <sub>EN</sub> )	8V
Power Dissipation	Internally Limited
ESD Rating, Note 3	2kV

## **Operating Ratings** (Note 2)

Supply Voltage (V <sub>IN</sub> )	1.4V to 6.5V
Bias Supply Voltage (V <sub>BIAS</sub> )	3V to 6.5V
Enable Input Voltage (V <sub>EN</sub> )	0V to V <sub>BIAS</sub>
Junction Temperature Range	–40°C ≤T <sub>J</sub> ≤ +125°C
Package Thermal Resistance	
S-PAK( $\theta_{JC}$ )	2°C/W

## **Electrical Characteristics**

$T = 0 \Gamma^{\circ} O \dots H h \lambda \lambda = \lambda \lambda \lambda = 0$	$4 \cdot (\cdot) \cdot (\cdot) = 1 \cdot (\cdot) \cdot (1 \cdot (1$	$d_{100} = 10^{\circ} O \times T \times 100^{\circ} O$	A unloss otherwise energified
$I_{AI} = Z_{0} \cup WIII V_{DIAC} = V_{OUT} + Z_{0}$	.1V; V <sub>IN</sub> = V <sub>OUT</sub> + 1V; <b>bold</b> values in	JICale -40 U < 1 < +125 U. <b>Note</b>	4: uniess otherwise specified.
			-,

Parameter	Conditions	Min	Тур	Max	Units
Output Voltage Accuracy	At 25°C, fixed voltage options	-1		+1	%
	Over temperature range	-2		+2	%
Line Regulation	V <sub>IN</sub> = 2.0V to 6.5V	-0.1	0.01	+0.1	%/V
Load Regulation	I <sub>L</sub> = 0mA to 3A		0.2	0.5	%
Dropout Voltage (V <sub>IN</sub> – V <sub>OUT</sub> )	I <sub>L</sub> = 1.5A		125	200	mV
	I <sub>L</sub> = 3A		280	400	mV
Dropout Voltage (V <sub>BIAS</sub> – V <sub>OUT</sub> ) Note 4	I <sub>L</sub> = 3A		1.5	2.1	V
Ground Pin Current, Note 5	I <sub>L</sub> = 0mA		25		mA
	I <sub>L</sub> = 3A		25	50	mA
Ground Pin Current in Shutdown	V <sub>EN</sub> ≤ 0.6V, (I <sub>BIAS</sub> + I <sub>CC</sub> ), <b>Note 6</b>		0.07	5	μA
Current thru V <sub>BIAS</sub>	I <sub>L</sub> = 0mA		20	35	mA
	I <sub>L</sub> = 3A		50	150	mA
Current Limit	V <sub>OUT</sub> = 0V		6.5	9	Α
Enable Input, Note 6		1			
Enable Input Threshold	Regulator enable	1.6			V
(Fixed Voltage only)	Regulator shutdown			0.6	V
Enable Pin Input Current			0.1	1.0	μA
Reference		1		•	
Reference Voltage	Adjustable option only	0.891 <b>0.882</b>	0.9	0.909 <b>0.918</b>	V V

Note 1. Exceeding the absolute maximum rating may damage the device.

Note 2. The device is not guaranteed to function outside its operating rating.

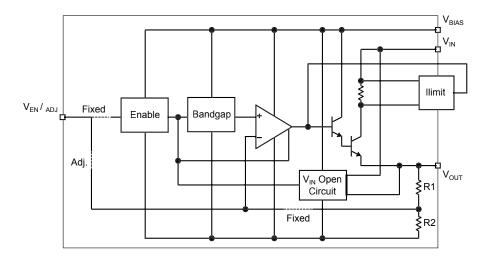
Note 3. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5k in series with 100pF.

Note 4. For  $V_{OUT} \leq 1V$ ,  $V_{BIAS}$  dropout specification does not apply due to a minimum 3V  $V_{BIAS}$  input.

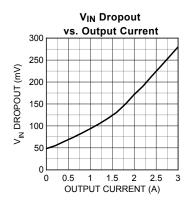
Note 5.  $I_{GND} = I_{BIAS} + (I_{IN} - I_{OUT})$ . At high loads, input current on  $V_{IN}$  will be less than the output current, due to drive current being supplied by  $V_{BIAS}$ .

**Note 6.** Fixed output voltage versions only.

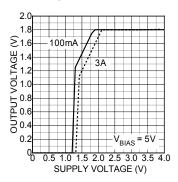
# **Functional Diagram**

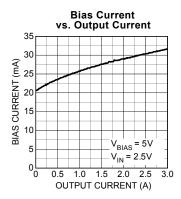


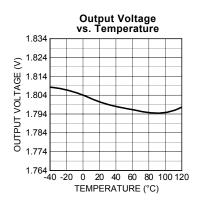
## **Typical Characteristics**

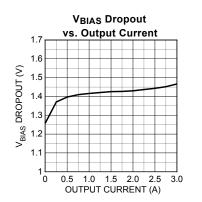


**Dropout Characteristics** 









Bias Current vs. Input Supply

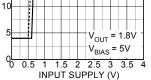
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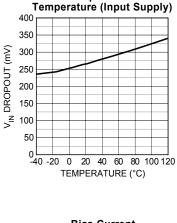
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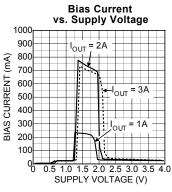
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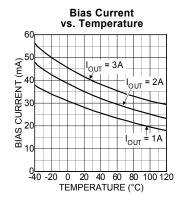
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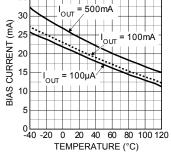


Dropout vs.





Bias Current vs. Temperature



Short Circuit vs. Temperature

### **Applications Information**

The MIC49300 is an ultra-high performance, low dropout linear regulator designed for high current applications requiring fast transient response. The MIC49300 utilizes two input supplies, significantly reducing dropout voltage, perfect for low-voltage, DC-to-DC conversion. The MIC49300 requires a minimum of external components and obtains a bandwidth of up to 10MHz. As a  $\mu$ Cap regulator, the output is tolerant of virtually any type of capacitor including ceramic and tantalum.

The MIC49300 regulator is fully protected from damage due to fault conditions, offering linear current limiting and thermal shutdown.

#### **Bias Supply Voltage**

V<sub>BIAS</sub>, requiring relatively light current, provides power to the control portion of the MIC49300. V<sub>BIAS</sub> requires approximately 33mA for a 1.5A load current. Dropout conditions require higher currents. Most of the biasing current is used to supply the base current to the pass transistor. This allows the pass element to be driven into saturation, reducing the dropout to 300mV at a 1.5A load current. Bypassing on the bias pin is recommended to improve performance of the regulator during line and load transients. Small ceramic capacitors from  $V_{BIAS}$  to ground help reduce high frequency noise from being injected into the control circuitry from the bias rail and are good design practice. Good bypass techniques typically include one larger capacitor such as a 1µF ceramic and smaller valued capacitors such as 0.01µF or 0.001µF in parallel with that larger capacitor to decouple the bias supply. The  $V_{BIAS}$ input voltage must be 1.6V above the output voltage with a minimum  $V_{BIAS}$  input voltage of 3V.

#### Input Supply Voltage

 $\rm V_{\rm IN}$  provides the high current to the collector of the pass transistor. The minimum input voltage is 1.4V, allowing conversion from low voltage supplies.

### **Output Capacitor**

The MIC49300 requires a minimum of output capacitance to maintain stability. However, proper capacitor selection is important to ensure desired transient response. The MIC49300 is specifically designed to be stable with virtually any capacitance value and ESR. A 1 $\mu$ F ceramic chip capacitor should satisfy most applications. Output capacitance can be increased without bound. See typical characteristics for examples of load transient response.

X7R dielectric ceramic capacitors are recommended because of their temperature performance. X7R-type capacitors change capacitance by 15% over their operating temperature range and are the most stable type of ceramic capacitors. Z5U and Y5V dielectric capacitors change value by as much as 50% and 60%, respectively, over their operating temperature ranges. To use a ceramic chip capacitor with Y5V dielectric, the value must be much higher than an X7R ceramic or a tantalum capacitor to ensure the same capacitance value over the operating temperature range. Tantalum capacitors have a very stable dielectric (10% over their operating temperature range) and can also be used with this device.

### Input Capacitor

An input capacitor of  $1\mu$ F or greater is recommended when the device is more than 4 inches away from the bulk supply capacitance, or when the supply is a battery. Small, surfacemount, ceramic chip capacitors can be used for the bypassing. The capacitor should be placed within 1" of the device for optimal performance. Larger values will help to improve ripple rejection by bypassing the input to the regulator, further improving the integrity of the output voltage.

#### Thermal Design

Linear regulators are simple to use. The most complicated design parameters to consider are thermal characteristics. Thermal design requires the following application-specific parameters:

- Maximum ambient temperature (T<sub>A</sub>)
- Output Current (I<sub>OUT</sub>)
- Output Voltage (V<sub>OUT</sub>)
- Input Voltage (VIN)
- Ground Current (I<sub>GND</sub>)

First, calculate the power dissipation of the regulator from these numbers and the device parameters from this datasheet.

$$P_D = V_{IN} \times I_{IN} + V_{BIAS} \times I_{BIAS} - V_{OUT} \times I_{OUT}$$

The input current will be less than the output current at high output currents as the load increases. The bias current is a sum of base drive and ground current. Ground current is constant over load current. Then the heat sink thermal resistance is determined with this formula:

$$\theta_{SA} = \left(\frac{T_{J(MAX)} \pm T_{A}}{P_{D} - \left(\theta_{JC} + \theta_{CS}\right)}\right)$$

The heat sink may be significantly reduced in applications where the maximum input voltage is known and large compared with the dropout voltage. Use a series input resistor to drop excessive voltage and distribute the heat between this resistor and the regulator. The low dropout properties of the MIC49300 allow significant reductions in regulator power dissipation and the associated heat sink without compromising performance. When this technique is employed, a capacitor of at least  $1\mu$ F is needed directly between the input and regulator ground. Refer to Application Note 9 for further details and examples on thermal design and heat sink specification.

#### **Minimum Load Current**

The MIC49300, unlike most other high current regulators, does not require a minimum load to maintain output voltage regulation.

#### **Power Sequencing**

There is no power sequencing requirement for  $\rm V_{IN}$  and  $\rm V_{BIAS},$  giving more flexibility to the user.

### Adjustable Regulator Design

The MIC49300 adjustable version allows programming the output voltage anywhere between 0.9Vand 5V. Two resistors are used. The resistor value between V<sub>OUT</sub> and the adjust pin should not exceed 1k $\Omega$ . Larger values can cause instability. The resistor values are calculated by:

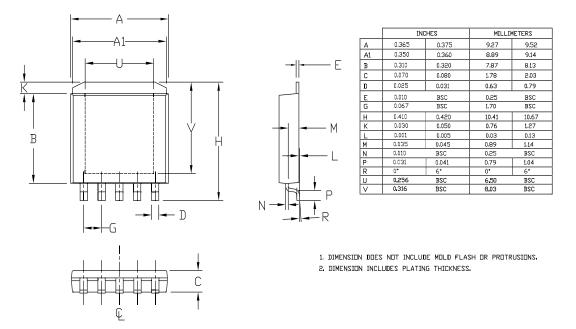
$$R1 = R2 \times \left(\frac{V_{OUT}}{0.9} - 1\right)$$

Where  $\mathrm{V}_{\mathrm{OUT}}\,$  is the desired output voltage.

### Enable

The fixed output voltage versions of the MIC49300 feature an active high enable input (EN) that allows on-off control of the regulator. Current drain reduces to "zero" when the device is sn}tñ◊wn, with only microamperes of leakage current. The EN input has TTL/CMOS compatible thresholds for simple logic interfacing. EN may be directly tied to V<sub>IN</sub> and pulled up to the maximum supply voltage

### **Package Information**



5-Pin S-Pak (R)

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